

Dynamic Bifurcations of the Active Magnetic Bearing-Rotor System

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Abstract: The Influence of nonlinear factors on the active magnetic bearing-rotor system (AMBR) with closed loop controller can't be derived from the classical linear control theories. It was analysed for the first time by the modern analytical methods used in nonlinear vibration. The Hopf bifurcations of AMBR at autonomous and non autonomous cases were researched by the Normal Form theorem and the improved averaging method respectively. Results show that the nonlinear factor of magnetism is one of the important factors causing the self-excited vibration of AMBR.

Key words: MBRS, Hopf bifurcation, Normal Form theorem

1. INTRODUCTION

Magnetic suspension is a new and high-performance mechano-electronic technique which uses controllable magnetic force to levitate workpieces (rotor, flatbed, etc.) without any contact. With no friction, no wear and high precision, low noise. It can be used even in many adverse circumstances such as high or super-low temperatures, vacuum, radiation environment. All these special advantages have opened up vast vistas for its application. Researchers in many countries are working efficiently to take the lead in this competition that may bring about a revolution in our modern industry and the future society.

The successful application of magnetic suspension in the superspeed train and enormous profits in industries have greatly encouraged the research in this field. International Magnetic Bearing Symposia were held once two years from 1988 in order to facilitate international inter-

change and cooperation. Meanwhile Magnetic Bearing Technique Conferences were held twice in America. Judging from the papers published, we can see that research is focused on key parts of the magnetic bearing, namely, sensors and magnetic executors, controllers in particular. The digital controllers are becoming much more promising. This paper deals mainly with the five-axis Active Magnetic Bearing rotor system (AMB) (Fig.1). As its radial bearing is of the popular eight-pole construction, the system chosen is representative. The nonlinearity of the magnetic force can make the dynamic model more precise. With this in mind we have investigated the influence of the parameters on its nonlinear vibration under the control of the P. D. controller. The results are conducive to the designing of both analog and digital magnetic bearings.

2. THE DYNAMIC EQUATIONS OF THE SYSTEM

2.1 The formula of the magnetic force of AMB

The formula of the magnetic force of the axial magnetic bearing can be given as follows:

$$F = K_F (G_0 + C_{10}u + C_{01}i + C_{20}u^2 + C_{11}ui + C_{02}i^2 + C_{30}u^3 + C_{21}u^2i + C_{12}ui^2 + C_{03}i^3) \quad (1)$$

Where K_F is the dimensionless coefficient of the magnetic force, u is the dimensionless displacement, $u = \beta u^* / g_0$, i is the dimensionless control electric current, $i = i_{CX} / I_0$, g_0 is the interspace of the magnetic field when the axis of the rotor is stable.

Without considering the effects of the magnetic flux-leakage and the magnetic-coupling, the radial magnetic force of the eight-pole AMB

can be expressed by equation (1).

2.2 The dynamic equations of the five-axis-controlled rotor

Neglecting the dynamic disequilibrium, gyroscopic effect and radial coupling, we obtain the dynamic equation of the rotor which is suspended wholly by magnetic force as follows:

$$m u = F_u - G_u + F_e \tag{2}$$

Where m:the mass of the rotor

u:the displacement component of the rotor at the suspended

point. In equ. (2),it can represent X,Y or Z direction.

F_u : the suspending force of the AMB

F_v : the perturbation force caused by the unequal mass

G_u : the static load of the rotor in u direction

2.3 The equation of the field-excitation electrocircuit and the output voltage of the controller (Fig.2)

Assuming that the controller, sensor, amplifier and field-excitation electrocircuit have the same construction, we have the equation of the field-excitation electrocircuit:

$$V_{cx} = R i_{cx} - \frac{d\psi_1}{dt} + \frac{d\psi_2}{dt} \tag{3}$$

$$\begin{cases} \psi_1 = \frac{\mu_0 N^2 A}{2(g_0 + \beta u^*)} (I_0 - i_{cx}) \\ \psi_2 = \frac{\mu_0 N^2 A}{2(g_0 - \beta u^*)} (I_0 I_r + i_{cx}) \end{cases} \tag{4}$$

The output voltage of the P.D. controller is

$$V_{cx} = \lambda K_p (u^* + K_d \frac{du^*}{dt}) \tag{5}$$

Select the appropriate value for I_0, I_r , then $K I G_0 = G_u$

2.4 The dynamic equation of the system

Having simplified eqs.(1)–(5), we obtain the dynamic equation of the system

$$\dot{x} = A x + f(\Omega\tau) + F(x) \tag{6}$$

Where

$$x = (u \frac{du}{d\tau} \frac{di}{d\tau})^T = (x_1, x_2, x_3)^T$$

$$A = \begin{vmatrix} 0 & 1 & 0 \\ C_{10} & 0 & C_{01} \\ a_2 & a_3 & a_4 \end{vmatrix}$$

$$f(\Omega\tau) = \begin{vmatrix} 0 \\ \wedge \sin(\Omega\tau) \\ 0 \end{vmatrix}$$

$$F(x) = \begin{vmatrix} 0 \\ (C_{20} x_1^2 + C_{11} x_1 x_3 + C_{02} x_3^2 + C_{30} x_1^3 \\ + C_{21} x_1^2 x_3 + C_{12} x_1 x_3^2 + C_{03} x_3^3) \\ b_0 x_1 x_2 - a_2 x_1^3 + 2x_1 x_2 x_3 \\ - a_4 x_1^2 x_2 + a_3 x_1^2 x_3 \end{vmatrix}$$

Take the accommodation coefficients of the control electrocircuit as the bifurcation parameters.

3. THE NONLINEAR VIBRATION OF THE SYSTEM

3.1 The autonomous case

The system can be regarded as autonomous when we investigate its axial activity or the radial activity while the center of the rotor is stable:

$$\dot{x} = A x + F(x) \tag{6.1}$$

when λ is at the critical value, the parameters of the system satisfy

$$a_2 + a_3 a_4 = 0 \tag{7}$$

The eigenvalues of matrix A are $\pm \omega_1 a_4$ ($a_4 < 0$), physical condition is naturally satisfied and ω satisfies $\omega^2 = -(c_{10} + a_3 a_4)$

If we denote the critical values of the parameters when $\lambda = \lambda_0$ by signs in the preceding equations, and take perturbation $\lambda = \lambda_0 + \mu$, we have $A(\mu) = A_0 + A_1 \mu$

where

$$A_0 = \begin{vmatrix} 0 & 1 & 0 \\ C_{10} & 0 & C_{01} \\ -a_3 a_4 & a_3 & a_4 \end{vmatrix}$$

$$A_1 = \begin{vmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ -k_p a_4 & -k_d a_4 & 0 \end{vmatrix}$$

Substituting transition $x = T * T / 0$ y into equ.(6.1), we have:

$$\dot{y} = J_0 y + J_1 \mu y + T_0^{-1} F(T_0 y) \quad (8)$$

where

$$C = \frac{a_4(a_4 k_p - k_d \omega^2)}{2 \det(T_0)},$$

$$d = -\frac{a_4 \omega(k_p + a_4 k_d)}{2 \det(T_0)},$$

$$e = \frac{a_4 \omega(k_p + a_4 k_d)}{2 \det(T_0)},$$

$$J_0 = \begin{vmatrix} 0 & -\omega & 0 \\ \omega & 0 & 0 \\ 0 & 0 & a_4 \end{vmatrix}$$

$$J_1 = \begin{vmatrix} d & -c & 0 \\ c & d & 0 \\ 0 & 0 & e \end{vmatrix}$$

$$T_0 = \begin{vmatrix} 1 & 0 & 1 \\ 0 & -\omega & a_4 \\ -\frac{c_{10} + \omega^2}{c_{01}} & 0 & \frac{a_4^2 - c_{10}}{c_{01}} \end{vmatrix}$$

Using the Center Manifold Theorem, we obtain the center manifold of the system(8):

$$y_3 = H_{20} y_1^2 + H_{11} y_1 y_2 + H_{02} y_2^2 \quad (9)$$

Substituting equ.(9) into the first two of equ.(8), we obtain the following equations of the center manifold:

$$\begin{cases} \dot{y}_1 = d\mu y_1 - (\omega + c\mu)y_2 + a_{20}y_1^2 + a_{11}y_1 y_2 \\ \quad + a_{02}y_2^2 + a_{30}y_1^3 + a_{21}y_1^2 y_2 + a_{12}y_1 y_2^2 + a_{03}y_2^3 \\ \dot{y}_2 = (\omega + c\mu)y_1 + d\mu y_2 + b_{20}y_1^2 + b_{11}y_1 y_2 \\ \quad + b_{02}y_2^2 + b_{30}y_1^3 + b_{21}y_1^2 y_2 + b_{12}y_1 y_2^2 + b_{03}y_2^3 \end{cases} \quad (10)$$

The Poincaré Birkhoff Normal Form of the system is

$$\begin{cases} \dot{r} = r(d\mu + ar^2) + h.o.t \\ \dot{\theta} = \omega + c\mu + br^2 + h.o.t \end{cases} \quad (11)$$

It can be easily seen that when $du/a > 0$, there will be the limit cycle in the system; when $a > 0$ the limit cycle is unstable, otherwise it is stable, and there will be self-excited vibration in the system.

3.2 The nonautonomous case

If we take the unequal mass of the rotor into consideration, the radial motion of the rotor is nonautonomous. Then let us study its resonance and disresonance, respectively.

a) Disresonance

Rewrite the equations of the system as ($u = \varepsilon \eta$)

$$\dot{x} = A_0 x + f(\Omega \tau) + \varepsilon [A_1 \eta x + F(x)] \quad (6.2)$$

From the orthogonality of the fundamental solutions of a equation and its adjoint equations, we can derive that $G(t)$ satisfies normalizing condition $G^*(\tau)G(\tau) = E$ (E is 3×3 identity matrix)

We also know that $x^*(\tau) = \frac{\wedge}{\omega^2 - \Omega^2} (\sin \Omega \tau \Omega \cos \Omega \tau a_3 \sin \Omega \tau)^T$ is the particular solution of equ. $\dot{X} = A_0 X + j(\Omega \tau)$ which is the linearizing equation of equ.(6.2). Then take transition

$$X = G(\tau)b + X^* \quad (12)$$

Substituting it into equ.(6.2), we obtain

$$\begin{aligned} b &= \varepsilon a^*(\tau) [A_1 \eta x + F(G(\tau)b + x^*)] \\ &= \varepsilon H(b, \omega \tau, \Omega \tau) \end{aligned} \quad (13)$$

$$\text{Let } \begin{cases} b_1 = r \cos \theta \\ b_2 = r \sin \theta \\ b_3 = B_3 e^{-a_4 \tau} \end{cases} \quad (14)$$

$$\text{then } \begin{cases} \dot{r} = \varepsilon [H_1 \cos \theta + H_2 \sin \theta] \\ \dot{\theta} = \frac{\varepsilon}{r} [-H_1 \sin \theta + H_2 \cos \theta] \\ \dot{B}_3 = a_4 B_3 + \varepsilon \tilde{H}_3 \end{cases}$$

Assuming the series solution of this equation is

$$B_3 = B_{32} + B_{33} + \dots \quad (15)$$

Where $B^{3j}(j=2,3,\dots)$ is the j -th homogeneous multinomial of $r \cos \omega \tau, r \sin \omega \tau, \cos \Omega \tau, \sin \Omega \tau$

Substituting it back into the first two of the

preceding equations, we get the dimension-reduced standard equations of equations(6.2):

$$\begin{cases} \dot{r} = \varepsilon\varphi = \varepsilon[\varphi_0 + \varepsilon\varphi_1 + \dots] \\ \dot{\theta} = \varepsilon\varphi^* = \varepsilon[\varphi_0^* + \varepsilon\varphi_1^* + \dots] \end{cases} \quad (16)$$

Where

$$\varphi_0 = \frac{ra_4c_{01}}{2(a_4^2 + \omega^2)} \{ (K_p + a_4K_D)\eta + (a_3 - 0.5a_5)(r^2 + \frac{a_3\Delta^2}{\omega^2 + \Omega^2}) \}$$

$$\varphi_0^* = 0.5\{ \frac{1}{\omega}(c_{30} + a_3c_{21} + a_3^2c_{12}) + (0.75r^2 - \frac{1.5\Delta^2}{\omega^2 - \Omega^2} + \frac{c_{01}a_4^2k_p\eta}{(a_4^2 + \omega^2)\omega}$$

$$- \frac{c_{01}\omega}{a_4^2 + \omega^2} [a_4k_D\eta + (a_3 - 0.5a_5)$$

$$(r^2 + \frac{\omega\Delta^2}{\omega^2 - \Omega^2}) \}$$

Using the averaging method, we can easily know that the linear approximate stationary solution satisfies

$$\begin{aligned} & \frac{0.5y_0c_{01}a_4}{a_4^2 + \omega^2} [(k_p + a_4k_D)\eta \\ & + \frac{\Delta^2a_3(a_3 - 0.5a_5)}{\omega^2 - \Omega^2} + (a_3 - 0.5a_5)y_0^2] = 0 \quad (17) \end{aligned}$$

This is the Hopf bifurcation equation of this system. Obviously, when

$$\frac{(k_p + a_4k_D)\eta}{a_3 - 0.5a_5} + \frac{a_3\Delta^2}{\omega^2 - \Omega^2} < 0,$$

the limit cycle exists, and when $a_3 - 0.5a_5 > 0$, the limit cycle is steady and self-excited vibration appears in the system. Otherwise the cycle is unsteady.

$$(\omega - \Omega = \varepsilon\delta, \delta < 0(1))$$

b) Main resonance

Rewrite the equation of the system as

$$\dot{x} = A_0x + \varepsilon[A_1\eta x + f(\Omega\tau) + F(x)] \quad (6.3)$$

Take transition $G = a(\tau) b$, and the equations(6.2b) can be transformed into

$$\begin{aligned} b &= \varepsilon G^*(\tau)[A_1\eta G(\tau)b + f(\Omega\tau) + F(G(\tau)b)] \\ &= \varepsilon H \end{aligned} \quad (18)$$

Then substituting equ.(13), we have

$$\begin{cases} \dot{r} = \varepsilon(H_1\cos\theta + H_2\sin\theta) \\ \dot{\theta} = -(\omega - \Omega) + \frac{\varepsilon}{r}(-H_1\sin\theta + H_2\cos\theta) \\ \dot{B}_3 = a_4b_3 + \varepsilon B(r, \theta, \Omega\tau) \end{cases}$$

Substituting it back into equ.(19), we have the dimension-reduced standard equations

$$\begin{cases} r = \varepsilon\varphi = \varepsilon[\varphi_0 + \varepsilon\varphi_1 + \dots] \\ \theta = \varepsilon\varphi^* = -(\omega - \Omega) + \varepsilon[\varphi_0^* + \varepsilon\varphi_1^* + \dots] \end{cases} \quad (20)$$

Then using the averaging method, we obtain the response equations of the resonant system:

$$\begin{cases} \frac{ya_4c_{01}}{a_4^2 + \omega^2} [(K_p + a_4k_D)\eta + y^2(a_3 - 0.5a_5)] + \frac{\Delta\cos y}{\omega} = 0 \\ \Omega - \omega + \frac{\Delta\sin y}{2\omega y} + 0.5\{ -\frac{a_4c_{31}\eta}{a_4^2 + \omega^2} (\omega k_D - \frac{a_4k_p}{\omega}) + \\ y^2[-\frac{c_{01}\omega}{a_4^2 + \omega^2} (a_3 - 0.5a_5) \\ + 0.75\omega(c_{30} + a_3c_{31} + a_3^2c_{12}) \} \} = 0 \end{cases} \quad (21)$$

4 CONCLUSIONS (Fig.1–Fig.4)

(1) The nonlinearity of the magnetic force is also one of the important causes that stimulate the self-excited vibration in the AMB.

(2) The rules of conditional stability: In the case of nonresonance, the system can work normally if any one of these two conditions is satisfied:

a) The amplitude of the unstable limit cycle is greater than g_0 .

b) The amplitude of the stable limit cycle is smaller than g_0 .

(3) The leappformance will appear in the system in the condition of resonance. But the resonance peak can be eliminated by changing proportionally the parameters of the system.

(4) The differentiate parameters of the control circuit have great effect on the function of the system.

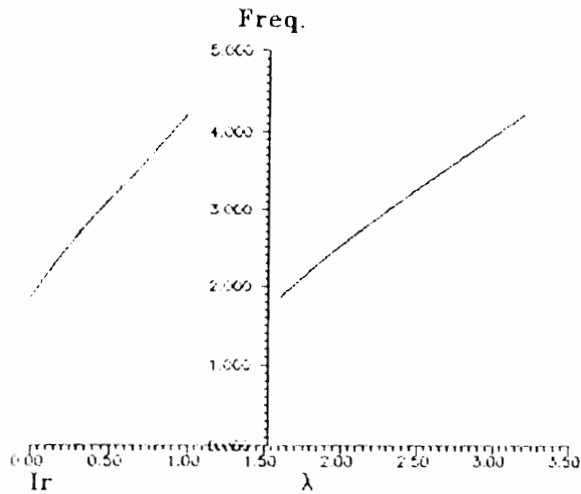


Fig.1 Ir's and λ 's effects on Freq

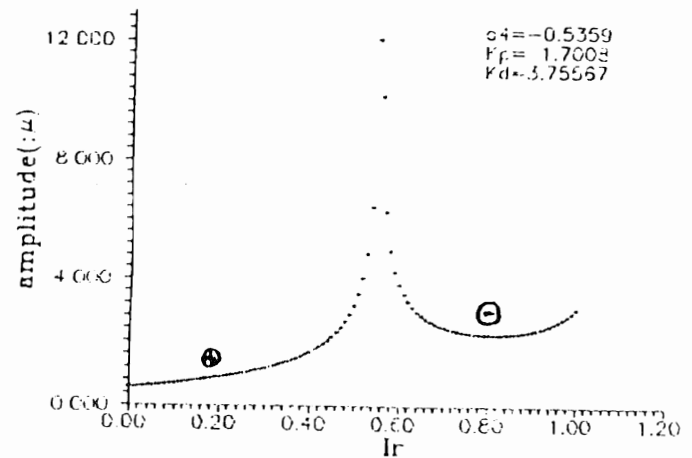


Fig.2 Ir's effects on amplitude

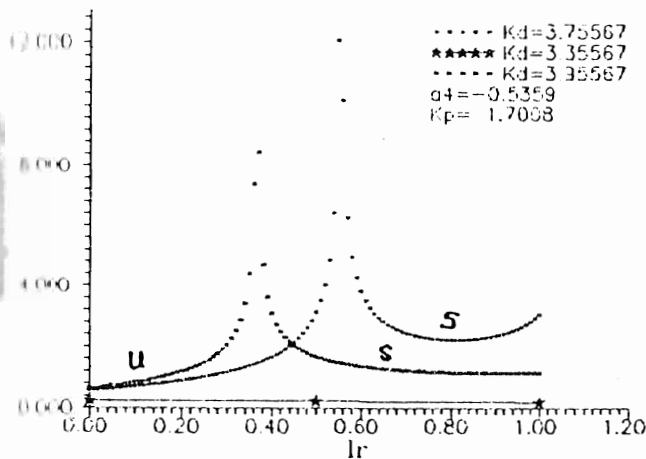


Fig.3 Ir's effects on amplitude

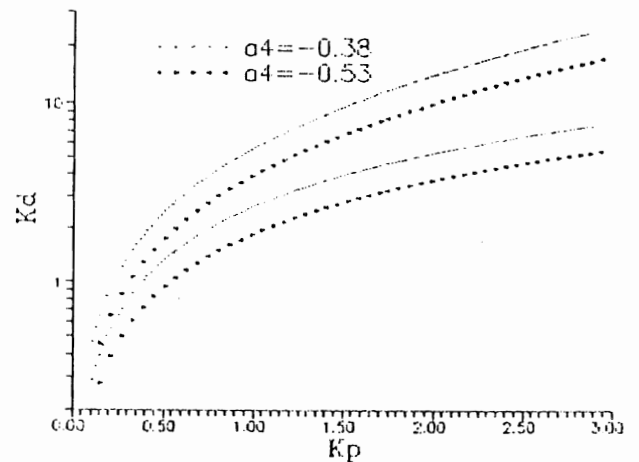


Fig.4 Kp's effects on K_D

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